

# Survey on Hardware Implementation of Random Number Generators on FPGA: Theories and Experiments Analysis

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**Abstract.** this paper introduce a survey on PRNG on FPGA...

**Keywords:** Random Number Generator, PRNG, TRNG, Chaotic PRNG, Cryptography, Security, FPGA

## 1 Introduction

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## 2 Random Number Generator: Theories and Classification

We can find many implementation of RNG in Software and Hardware but all can be classified generally as PRNG, TRNG and Hybrid, but new concept has been introduced this last year's defined by parallel and chaotic generator.

### 2.1 Pseudorandom Galois Generators

$F_2$ -linear generators are a well-known PRNG class of and a special case of matrix linear recurrence modulo 2, where they are appropriate for low power and high speed requirement. However with the limitation of the shift register state, these generators enter to a finite and repeated state cycle and have a short period cycle. Because of that many hardware optimizations are proposed to perform a good random and increase the period. A linear Random Number generator is defined following two equations, where the first equation (1)  $x_t = (x_{t,0}, \dots, x_{t,k-1})^s \in F_2^k$  is the  $k$ -bit state vector at step  $t$ . The second equation (2)  $y_t = (y_{t,0}, \dots, y_{t,w-1})^s \in F_2^w$  is the  $w$ -bit output vector at step  $t$ ,  $k$  and  $w$  are positive integers. Where,  $A$  is a  $k \times k$  transition matrix with elements in  $F_2$ ,  $B$  is a  $w \times k$  output transformation matrix with elements in  $F_2$ , and  $u_i \in [0, 1]$  is the output at step  $t$ :

$$x_t = A * x_{t-1} \quad (1)$$

$$y_t = B * x_t \quad (2)$$

$$u_t = \sum_{\ell=1}^w y_{t,\ell-1} 2^{-\ell} = y_{t,0} y_{t,1} y_{t,2} \quad (3)$$

The characteristic polynomial of the matrix A is  $x_t = a_i x_t - 1 + \dots + a_k x_{t-k}$

## 2.2 Discrete Dynamics

## 3 Pseudo Random Number Generator

Some of the algorithms that generate pseudo random numbers are Blum Blum Shub, Inversive congruential generator, ISAAC (cipher), Lagged Fibonacci generator, Linear congruential generator, Linear feedback shift register, Mersenne twister, generalized feedback shift register (GFSR), twisted GFSR (TGFSR), Multiply-with-carry, Well-Equidistributed-Long-period Linear, Xorshift and Cellular Automata. As well as separately implementation of these structures, with use one or several of them, hybrid PRNGs can be designed. An important property of all these generators is that they are special cases of a general class of generators whose state evolves according to a (matrix) linear recurrence modulo 2 and the bits that form the output are also determined by a linear transformation modulo 2 applied to the state.

### 3.1 F2-Linear Generator

*Linear Feedback Shift Register Generators* or LFSR use a feedback polynomial driven by an Exclusive-OR (XOR) coefficient during a period of  $2^n - 1$  and an initial input called the seed. Many simple implementation can be found but few of them has an optimization version as in this work [15], where presenting two PRNG based only on LFSR. The first is called Shrinking Generator (SG) using 2 LFSRs of 67-bit and the generated outputs are from the second LFSR-2 if the first LFSR-1 is "1" or put "0" with a period of  $(2^{L_2-1}) * 2^{L_1-1}$ . When, in the second PRNG is called Alternating Step Generator (ASG) that use a third LFSR-3 of 141-bits to control which output will be taken of the two LFSR of 131/137bit and with a period of  $2^{L_1}(2^{L_2-1})(2^{L_3-1})$ .

*linear Congruential Generators* or LCG are another linear recurrence decrypted by  $x_{i+1} = (ax_i + b) \text{ mod } 2^n$  where "a" is called the multiplier, "b" the increment, and "m" the modulus. For Hardware implementation we can found in [22] proposing two version, the first version is a coupled of two linear congruential generator (CLCG), where the LCG-1 generate a control bit to select the output bit from the second LCG-2 following this equation  $x_{t+1} = (a_1 x_t + b_1) \text{ mod } 2^n$ . However in the second version is a pipeline of two CLCG of 4 stages to generate a 4-bit simultaneous, where in every level will compute a n-bit/n-pipeline with two n-bit additions following this equation  $x_{t+1} = (2^r * x_t \text{ mod } 2^n) + (x_i + b_1) \text{ mod } 2^n$ , and the result of every stage will be feedback and processed in the earlier and following stage.

*Gaussian Generators* Flowing thos definition, an series of optimised versions was developped in [37]. where the authors

However, in [21], the authors present an alternative implementation of the gaussian PRNG based on digital decimator and LFSR, where the decimator will use the output binary of the LFSR to generate white noise whose auto-correlation as input of the central limit theorem (CLT)to generate a Gaussian-distributed pseudorandom number.

*LUT and Accumulator Generators* are an digital component used or implemented on FPGA to accelerate optimization, where Lookup-Tables are FPGA depended technologies resource and accumulator is a combinatory logic that use FPGA resources. Some of works based on LUT is in [40], [38] and [39], the authors present a serie of PRNG based F2 Linear matrix recursive algorithms using FPGA resources as Flip-Flips, Lookup-Tables, Shift Registers, and bloks of RAMs. The basic idea is to provide a maximum area efficiency using this optimization, where for each row of the recurrence matrix A can be maped as a XOR gate to implement it in a single input LUT and testing the characteristic polynomial of each matrix for primitivity. However, the big drawback of this method is that to create a long period sequence, a large number of LUT-FF pairs must be used. Even if an application only needs 64 bits per cycle, it must use 512 LUT-FFs to get a period of  $2^{512-1}$ , so 87 percent the performance is wasted. Where in [17], The authors a PRNG of 9 stage of 8 bit digital accumulator. The input of the generator is the quantisation error mapping function of the previews stage and the output is feedback to earlier stage. However the Accumulator coefficients inputs is varying in time using an LFSR PRNG.

*Blum Blum Shub Generators* or BBS are an efficient cryptographic secure PRNG proposed [1986] to solve the quadratic residue problem [1999] represented by the Rabin function of  $x_t = x_{t-1}^2 \bmod n$  where n is the product of two Blum primes and congruent to(3 modulo 4). However the output bit generated by extracting the least significant bits  $j = c * \log(\log(n))$  of  $x_t$ , where c is a constant and then added to the generated binary sequence  $y_t = x_t^2 \bmod j$ . Although, few work using BBS has been found implemented on FPGA, where in [34] the authors present a comparison between a 4-bits LFSR and a 16-bits BBS PRNG in terms of area/speed, but their results are without any optimization. However, in [32] the authors present an implementation comparison of BBS OF 160/512 bits using different algorithm for modular multiplication ( $A * B \bmod M$ ) for RFID tag applications. The algorithms compared are classical combinational multiplier, a classical shift adder and multiplication, a Karatsuba's multiplication and Montgomery's direct modular multiplication. However, they limited to just in the area results and they claim the Montgomery iterative approach is the low cost area.

*Mersenne Twister Generators*or MT are a word-wise recurrence instead of bit-wise of LFSR, where the first implementation use a generalized FSR "GFSR" and the output will be a state vector or a array  $x[t] = (x_{M1} \otimes x_{M2} \otimes \dots)$  where M is the middle word  $1 < M < t$ . However, GFSR are a know to not reach the maximum period and to resolve that, a matrix recurrence twisted GFSR

"TGSSR" has been proposed to increase the period with a feedback polynomial of  $x_{t+k} = [x_{M+k} \otimes (x_k^u | x_{k+1}^l) .A]$  ( $k=1,2,3$ ), where  $x[0]^u, x[1]^l$  are the less and most significant bit of  $x_t$  ( $u,l$  are tempering bit shifts). The outputs are tempered by a bitwise multiplication using binary matrix  $y = x_t T$  to uniform the outputs and also perform the statistical properties by reducing the dimensionality of equidistribution. For a long period, the Mersenne twister is a special case of TGFSR, where MT11213 has a period of  $2^{19937-1}$  and MT19937 will have  $2^{19937-1}$ .

Some implementation of TGFSR are in [2], where the authors compare two implement version of TGFSR, where the first resumed by the basic Mersenne Twister PRNG (MT19937, MT11213) using Generalised Feedback Shift Registers (GFSRs). Then, the two optimised TGFSR version "Ran" and "Ranq1" of TGFSR has been proposed but based just on multiplication of fixed precision integers (with overflow) for area optimization unlike of the first version that use block memories RAMs for multiplications.

As for Mersenne Twister, we can find in [45] where the authors propose a 3 stages pipeline version of MT19937 that use a Dual-port Bram memories of the FPGA to generate a long period even it take 3 cycle/sample. However, in [27] the authors present a 3R/1W RAM structure resulting a single sample per cycle. Another implementation in [4] presenting a parallel MT19937 using single block RAMs and in [41] for MT2203 using multi ported block of RAMs to reduce HW resource and to store the state vector. Where, in [11] the authors propose two parallel implementation of MT19937 by reducing the IO ports and bank memories, where the Interleaved Parallelization generate each bit by a P memories bank separately during a period of  $T = (P/N_{bits}) * (N_{IO}/clock)$ , when the Chunked Parallelization is based on the idea the output of each bank can be re-used as the far recurrence input for the following bank using fewer block RAMs than the same degree of IP.

Finally for a recent reseache in [14], the authors present two MT PRNGs based on the same set of parameters of MT19937 where the first use RAM banks memories storage element, the second MT is designed to use circular buffer as an alternative solution. The new solution is based on the fixed relationship between the indexes of the words and considering that each step of the recurrence  $x_t$  is replaced by  $x_{t+k}$  in the work area. This way, the linear recurrence and the buffer of registers can be considered as a circular buffer where the linear recurrence is carried out by some combinational logic between the input and the output of the buffer. Hence the architecture is simplified as no logic for the table indexes is needed.

### 3.2 Cellular Automata

Cellular Automata (CA) are proposed as a Class 3 for exhibit chaotic or pseudo-random behavior by Wolfram 'A New Kind of Science' and can be represented as an array of cells that can hold and update their state dependent on local rules and the state of their neighborhood. The basic CA can combine 3 cells where each of them has two possibilities resumming a state machine with 8 ( $2^3$ )

possibilities binary configuration for each CA and resulting a 256 ( $2^8$ ) possible rules generally referred to by their Wolfram code. If the rule of a CA involves only XOR logic, then it is called a linear rule, where it will be referred to as complement rules when using XNOR logic. Otherwise, a linear CA represent all cells that has linear rules and an additive CA if they combine the linear and complement rules. another type which is the uniform CA that has the same rule for all cells, else it is a hybrid CA. Finally, a null boundary CA is when both the left and right neighborhood of the leftmost and rightmost terminal cell is connected to logic 0-state.

In 2003 [CA-1] Tkacik implemented on a custom IC that combines the outputs of a hybrid 90/150 rules an 37 bit CA was combined with a 43-bit LFSR to produce a maximum length RNGR. however and to pass pass all the DIEHARD tests, it was found that the LFSR and CA must be clocked at different frequencies to create a random output sequence. Then and to resolve Clocking issue or [CA-1], in [3] propose a solution by XORing the last bit of HCA with the last bit of LFSR to generate 1-bit per clock cycle, and they found the best combination for a high quality of PRNG is 37-bit LFSR with 16-bit CA. Where, in [7] they compare the preview work LFSR/HCA [3] with an implementation of a SPCA [18] that use 90/156 rules, and they find the SPCA give a better throughput than the LFSR/HCA even it fail in one statistic test. Also, in [33] propose another combination circuit of CA and NLFSR block based on A2U2 design to resist more to various forms of cryptanalysis, such as correlation attacks and algebraic attacks and the output will pass to a mixer mechanism to increase the complexity for decryption.

When, others combination using only HCA like in [1] proposing a combination of two Hybrid HCA as a PRNG and block cipher for an encryption system. The first HCA-1 use two rules 90/150 as a real-time key stream generator and the second HCA-1 use 51/153/195 rules. however, To select witch rules will be used by the block cipher HCA-2, the PRNG or HCA-1 generate an encryption rules to switch the rules and provide each cell of HCA-2 is own rules. Where, in [13] they create an automatic software tool based on Algebraic Normal Form (ANF) representation to generate an RTL code of any hybrid CA depending on ID rules. The results show that by using ANF representation, the output ( $y = [K_0 \otimes K_1(U_1) \otimes K_2(U_2) \otimes K_3(U_3) \otimes K_4(U_1 * U_2) \otimes \dots K_7(U_1 * U_2 * U_3)] \otimes *mask$ ) with ID=101 and 3 neighbor are identical to Matlab results. Then in [20], they claim that by using a chain of HCA instead of single HCA will increase the ratio of frequency/ares and cryptography.

Another solution is proposed in [25] using a new way to implement the rules for 1-d CA using the real time computer clock sequence, where the initial state configuration and the length is the product of day, month, year, hour, min and seconds. however and over a period of  $[t = x(60 - x)]$ , the functional rules as first rule is the from the product of minutes by seconds and the second rule is the number of minutes divided by the number of seconds multiplied by a constant.

### 3.3 Chaotic Random Number Generator

The interests of using the chaos theory to generate random number has been increase due to the sensitivity to initial conditions, unpredictability and ability to reciprocal synchronization. however, the finite precision of arithmetic and quantization generate an non-ideal and periodic random number. We can found chaotic RNG generated by a differential equations in a continuous time domain, or the most implemented is generated by a recurrent sequences using chaotic maps in a discrete time domain  $x^0 \in R$  where  $x^{n+1} = f(x^n)$ . Hence, from the cryptographers point of view, the most essential drawbacks of the chaotic systems are: weak resistance to recovery of parameters, relatively low digital precision comparing to the analog generators, periodicity problems, non ideal probability distributions, higher level of correlation and suffer from serious dynamical degradation due to quantization error and finite representation of system states, including loss of periodicity and shorter pseudo-orbits.

*Chaotic Map Generators* describe here are generators based on a polynomial mapping that is equivalent to recurrence matrix degree 2, where for chaotic map generators they use a non-linear dynamic transformation. For this survey, we can find a logistic map based on this equation of  $x_{t+1} = rx_t(1-x_t)$  where  $r$  is the biotic potential ( $3,57 < r < 4,0$ ). Also, there is Hnon map which is also a discrete-time dynamical system according to the equations of  $x_{t+1} = y_t + (1 - ax_t^2)$  and  $y_{t+1} = bx_{t+1}$  where  $a$  and  $b$  are canonical parameters. Following that and based on earlier study of those different chaos map system implemented on FPGA in [8] and [9], the authors implement two version of PRNG based on chaotic logistic map using XSG Xilinx tool illustrated in [10]. Where The first is log-LUT based on on LUT blocs and high-speed carry line of the FPGA and the second is Log-DSP that use directly DSP of FPGA. However, they add delays to ensure parallel sequence generation and a complex initial sequence used for a better NIST test results.

Another work presented in [31], where the authors demonstrate a mixed version using tow chaotic map to allow increasing of the security level against plaintext attacks. They show that by coupling a chaotic encryption system (ENS) based on 2-D Hnon map used to generate the chaotic sequence, and control system (CRS) based on 1-D logistic map to control a multiplexer to choose the output of ENS according to the value generated by the logistic map by XORing the MSB of 32-bit of CSR with his it neighbor LSB following some rules. In the same approach and in [19] the authors use chaotic logistic map to generate output for each period increased by a Reseeding module where the output sequence will be Xored with a vector mixing based on auxiliary linear generator. Where, in [46] present a chaos system based on three Dynamic nonlinear transform arithmetic DNT process in a parallel structure that transform input 3 times and improve a high cycle-length and distribution output sequence, and that of each DNT module initiate his 2x256 code book and obtain the binary input sequence, then transformed and look up it using inputs as parameters  $x_{t+1} = x_t (C(w) R(q))$ .

*Spatio-temporal Chaos Generators* are a temporally chaotic dynamical system as chaotic map but also a spatially and there is many mathematical models can

be use to represent this type of generator. Where in [30] and to achieve a high operation speed, a bi-directional coupled chaotic map lattices (CML) has been used as a models represented by  $x_{t+1}(i) = (1 - \epsilon)f(x_t(i)) + \frac{\epsilon}{2}(f(x_t(i-1)) + f(x_t(i+1)))$  where n and i are respectively temporal and spatial indexes of discrete lattices,  $\epsilon$  is the couple coefficient, L is the number of the total spatial lattices and  $f(x)$  is a logistic map. They first deal with continuous domain with digitized all operand to be suited for HW implementation by using and modifying the CML to a finite integer set, then second and to avoid finite precision chaotic map problem, they compute only the insignificant bit is subject to output.

*Fibonacci post-processing Generators* is presented in [29] illustrating a non-autonomous 4-D hyperchaotic-based PRNG post processing based on Fibonacci series and driven by a 256-bit LFSR. The post processing is based on two loop feedback, where the first use a fixed 1-bit static rotation to suppress the short-term predictability and the resulting output after C cycles at the m-th bit position is . The second loop is based on a variable rotation controlled using Fibonacci series of K-bit to enhances differential sensitivity.

*Chaotic Iteration*

## 4 True Random Number Generator

*Phase-Locked Loop* or PLL circuit in general is derived by an external clock generator source like quartz or RC circuit, and wich can set static or dynamic configuration. A PLL is completely depended of the physical environment like power, temperature or others that cause a very high secure RNG and reduce attacks. the TRNG generator based PLL will use a jitter extractor to generate randomness, wich is a short-term variation of the clock propagation. The most common jitter measurements used by FPGA vendors are period jitter and cycle-to-cycle jitter. The period jitter is defined as the difference between the n-th clock period and the mean clock period. However and in FPGA, a PLL is is based on the size of the clock jitter, the frequency divided by the VCO and the his loop filter bandwidth. Generally, we can find an analog PLL TRNG that extract the "intrinsic jitter" causes by the nois of his VCO for exemple or a digital PLL generator that use a synchronous/asynchronous Flip-Flop as an extractor. we can find an old implementation in [16], where the authors propose an analysis about extracting randomness from the jitter of an PLL implemented on Altera FPLD. Their studies is based on detecting the jitter by the sampling the reference clock signal ( $F_{CLK}$ ) using a correlated signal synthesized in the PLL ( $F_{CLG}$ ) where  $F_{CLG} = F_{CLK}(K_M/K_D)$ , and the maximum distance between the two clock (CLK,CLG) must be minimum  $MAX(\Delta T_{min}) < \sigma_{jit}$ . However, they confirm in ideal environment condition and without a jitter the sampled output or random is deterministic under a period of  $TQ = K_D T_{CLK} = K_m T_{CLG}$ . Then, they conclude in a real condition  $\sigma_{jit} \neq 0$  the randomness is not deterministic and depending on jitter distribution where the  $MAX(\Delta T_{min}) = T_{CLK} * GCD(2K_M, K_D)/4K_M$ .

Where, in [35] the authors demonstrate by taking [16] as model and combined more than one PLL in parallel or series to increase the significantly sensitivity on the jitter  $S = F_{CLK}MAX(\Delta T_{min})$  and the output-bit of the generator compared to the use of one PLL. The configuration of multiple PLL are based on input/output length, VCO frequency and MUL/DIV factors ( $K_M/K_D$ ). In [36] the authors test the impact of the the change on operation condition environment as temperature of an PLL and illustrate that with low bandwidth of PFF cause a higher number of the critical samples, decreases the output jitter and thus increase the tracking jitter. As an PLL application system, the work proposed in [43] they explore an embedded system with TRNG and based on PLL to extract randomness from the jitter and propose two version where the slower of 40kbps can pass the statistic tests.

*ring oxialltor* In [24] and in [23], the authors propose a TRNG based on two ring oscillators clocked by different clock generated by an internal PLL on FPGA. the authors also extract the jitter of the 2 RO implement in only one CLB slice using a simpler. Where, in [12], propose a new approach that can replace RO based on inverters using XOR combination between Fibonacci (FIRO) and Galois ring oscillators (GARO). the main key consists of a number of inverters and connected in a cascade together with XOR logic gates forming a feedback in an analogous way where the feedback polynomial form is  $f(x) = (1+x)h(x)$  where  $h(1) = 1$  and the result show with the new method can achieve a stable state less than classical RO.

In [42], the authors propose a Hybrid implementation on FPGA of TRNG based on RO and PRNG based on BBS generators and with high operation frequency of 400Mhz. However, they generate a low off-chip frequency based on resistor and capacitors RC and it was notice by implementing BBS using ALU structure for squaring and modulo operation the period will be  $[(4.5 * n^2 + n)]$ .

*Self-timed ring STR* In [6] and [5], the authors propose another alternative more based on Self-Timed Ring (STR) robust to environment (power, temperature) than RO based inverter. The SRT approach consist of a ripple of L stage of FIFO as a ring  $(C_i)_{1 \leq i \leq L}$  with a phase of  $\Delta\varphi = T/2L$ , and extract jitter of each oscillator stage using two asynchronous handshaking protocol as even that can be "taken" or "bubble". However, the outputs randomness bits event  $(S_i)_{1 \leq i \leq L}$  will be samples using a flip-flop by the main clock and the result will be combined with a XOR operation  $\psi = s_1 \oplus s_2 \oplus \dots \oplus s_L$ . Secondly, the authors suggest that to avoid the limitation frequency of the STR by the long period delay, the maximum frequency is achieve when the propagation delay (forward and reverse static delay) is near to ring accuracy (N of token and bubble) and  $[N_T/N_B \cong D_{ff}/D_{rr} \simeq 1]$ .

*Metastability* In [44], the authors present a studies of using Metastability phenomena as a entropy source generated by 5 IRO stage. They claim that by implementing the inverter as loop ring and using a Control Clock Generator to switch the connectivity between the IRO stages flowing two mode (MS, Generation), the output voltage converges to metastability level and stays longer than using bi-stable circuit (Flip-Flop) causing a high entropy. However, the authors



wan to estimate the robustness of the system after applying the sampling process in a different process and environment variation modes using CMOS process and FPGA, and they find that it must added another stage for a higher quality output as decreasing the operation rate, applying a Von-Neumann post-processing and influences the loads (RC parasitic) of the last inverter, when it was noted that just post-processing is used in FPGA.

In [28], The authors propose a TRNG using the metastability of the flip-flop when there is a violation in setup/hold time. the system is based on closed-loop feedback mechanism for auto-adjustment on delay  $\Delta$  controlled by the Programmable delay lines (PDLs) stage based on LUT to avoid violation and maintain the metastability,, however the system use at-speed monitor to keep tracking the output bit probability and proportional-integral (PI) controller to decides to add/subtract the delay difference ( $\Delta \rightarrow 0$ ). The probability of the output is  $ProbOut = 1 = Q(\Delta/sigma)$  where  $Q(x) = 1/\sqrt{4\Pi} \int_x^\infty \exp(-u^2/2)du$ , where the updated/corrected delay difference is the difference between the bias/skew caused by the routing asymmetric with the delay induce by the environment condition and the correct delay injected by the PDL ( $\Delta = \Delta_p + \Delta_b - \Delta_f$ ). A revision version proposed by [26], to analyze the probability and maintain metastability state for a long period to avoid the deterministic state. however they use an extract hardware resource as memory for storing the outputs and use Hamming weight to calculate the probability bits histories.

## 5 Rndom Number Generator:Experiments Analysis

### 5.1 Statistic Test

*National Institute of Standrs and technologies* or **NIST** is based on hypothesis testing and includes a 15 batteries of mathematical and physical properties tests for RNG. It was developed to test the randomness of (arbitrary long) binary sequences produced by either HW/SW based cryptographic random or pseudo random bit generators with a fixed input parameters as sequence length of ( $10^3 < N < 10^7$ ) and ( $M < 55$ ) of binary sequences (sample size). For each  $N$  sequence produced, the NIST determine by prbability computed by  $P - value$  (level of test) of all different type of non-randomness exist, where the  $P - value$  must be more than a significance level  $\alpha[0.0001, 0.01]$  to classify it as a good RNG and equal to 1 to have perfect randomness. Practically, the NIST test evaluate in first the range of acceptable proportions of binary sequences passing the statistical test, then in second it evaluate the uniformity of the test sequence and computing on the basis of  $x^2$  test to the  $P - value$  obtained.

*TestU01* is now the most complete and difficult batterie of test of RNG. it was developed by Pierre L'Ecuyer and was implemented in the ANSI C languages with more than 516 tests resumed in 7 big batteries ( $P - value[0.001, 0.999]$ ). This new Battery of tests covers divers classical tests of the other batteries with new algorithm performance and cryptographic tests. Six predefined batteries of tests are already included, where the last batterie **FIPS** is the recall for NIST tests. The first three batterie (319 tests) are for sequences of random numbers

**SmallCrush**, **Crush** and **BigCrush**, and the three others ( 181 tests) are for bit sequences **Rabbit**, **Alphabit** and **Pseudo-DieHARD** designed to test a finite sequence contained in a random bits.

The other test batteries are the **Diehard**, it's include 18 test of Randomness and was developed by George Marsaglia. It was supposed to give a better way of analysis in comparison to original **FIPS** (16 tests) statistical tests. However and unlike NIST, the  $P - value$  have to belong to some fixed chosen interval  $[0 + \alpha, 1 - \alpha]$  with a level of signification  $\alpha$  of 5% for example. Where the **ENT** battery tests was developed to sequences of bytes stored in files and reports the results of those tests as Entropy and  $x^2$  tests.

The results of this survey presented by the authors, show some different level of what it concerned testing statistic of the RNG implementation from theories to practical results. Where the tests standard evaluate and updated to covert more the characteristic of the RNG starting by the simple to a complex and hard batteries of test. We can analysis the result by two category that it seen more stable and trust tests as the NIST and TESTU01, where the other batteries are even for old tests or they are included in those two batteries.

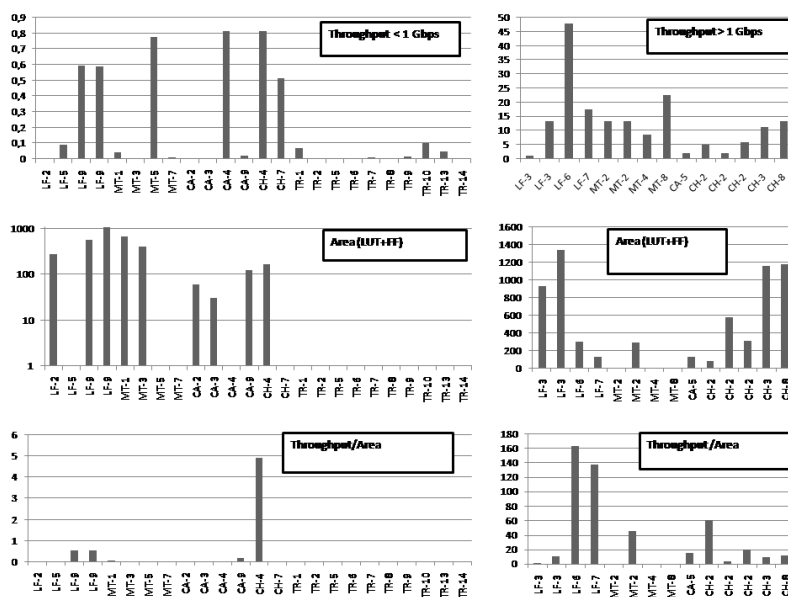
The PRNG analysis table show that not all papers has pass the NIST test, where Chaotic PRNG are th most success implementation of FPGA using different meethods. However, all work that use FPGA opetimised resourcec as LUT or accumulator can pass the test where others not. give a details about this section of tests, where there is some difficult to say it pass it or now specially the coverege of tests.

RNG	DieHard	FIPS	NIST	TestU01	AIS
PRNG	LF[4,6]	CHO[7]	LF[1*,3,5]	LF[4*,6*,7*,8*]	
	MT[1,3,8,7]		CA[5-9]	MT[1*,3*,8*,6**]	
	CA[2,3,5,8]		CHO[1,3,4,6,7,8]		
TRNG	IRO[7,8]	PLL[4]	PLL[1,2,5]	IRO[7**]	STR[10]
	STR[10]	STR[10,12]	IRO[6,8]		MTS[13]
		MTS[13]	STR[10,12]		
		MTS[14,15]			

Cryptography Secure

## 5.2 Hardware Implementation

Hardware Implementation



## 6 Conclusion

Conclusion

## 7 References

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